

## Multi-beam formation using a parasitic patch integrated microstrip array for 5G communications

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### Abstract

This work presents the design and realization of a compact multi beam microstrip antenna array intended for 5G communication at 28 GHz. The array is built around a miniature square patch element measuring  $2.95 \times 2.95$  mm $^2$ , integrated with a transformer-coupled feed network that enables multi-beam operation through controlled phase excitation at the input ports. A dual-port orthogonal feeding scheme is initially implemented to ensure low mutual coupling, with correlation coefficient and diversity gain analysis confirming strong port isolation and independent beam control. The concept is further expanded into a five-element array supplied by eight input ports, where the central element is parasitically excited to enhance boresight directivity. By adjusting the relative phase shift among the ports, the array can generate single, dual, triple, quad, and five simultaneous beams. Simulation results indicate that a single beam yields the highest gain and narrowest beam width, while increasing the number of beams leads to moderate gain reduction and broader half-power beam width. The prototype, fabricated on a Rogers RT5880LZ substrate ( $\epsilon_r = 1.96$ ), occupies a compact area of  $27.7 \times 27.7$  mm $^2$ . Performance validation shows a good match between measurements and simulations, with the measured resonance at 27.8 GHz closely aligning with the simulated 27.6 GHz. Experimental characterization was carried out using a Rohde & Schwarz ZVA 40 VNA and radiation pattern measurements were performed in an anechoic chamber capable of operation up to 70 GHz.

**Keywords:** fifth generation (5G), microstrip antenna array, multiple beams, parasitic patch, radiation pattern, rogers substrate.

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### 1. Introduction

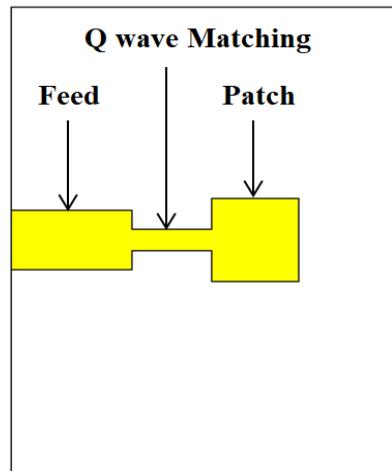
5G communication systems operate across an extensive portion of the radio spectrum, particularly within the millimetre-wave region. The FCC has allocated the 27.5–28.35 GHz band for 5G use, while the ITU identifies several broader ranges for global deployment, including 3.4–3.6 GHz, 5–6 GHz, 24.25–27.5 GHz, 37–40.5 GHz, and 66–76 GHz [1]. At these elevated frequencies, electromagnetic signals experience substantial free-space path loss and are easily obstructed, resulting in a reduced signal-to-interference-plus-noise ratio (SINR) [2]. To mitigate these effects, high-gain and highly directive antennas are essential for focusing energy toward intended directions. 5G New Radio (NR) divides its operating spectrum into Frequency Range 1 (FR1) below 6 GHz—covering many legacy LTE bands and Frequency Range 2 (FR2) above 24 GHz, which enables very high data throughput over

shorter distances. Prominent 5G NR bands such as n78 (3.5 GHz), n79 (4.7 GHz), n41 (2.5 GHz), and n28 (700 MHz) are widely adopted across global networks [3].

Several frequency bands, including 3.4–3.8 GHz and 25.5–27.0 GHz, have also been allocated for preliminary 5G trials and deployment. As the latest generation following 1G through 4G, 5G establishes a highly flexible and interconnected communication environment capable of supporting massive numbers of devices with improved reliability and user experience [4]. The transition from 4G to 5G promises significant enhancements, with data rates rising from roughly 35 Mbps to more than 50 Mbps, and latency decreasing from nearly 50 ms to about 1 ms. Operating across both low-frequency (<6 GHz) and high-frequency (6–100 GHz) ranges, 5G systems rely heavily on antenna arrays due to their ability to deliver strong directivity, higher gain, and precise beam steering features crucial for modern wireless infrastructure, radar systems, satellites, and advanced 5G applications [5-6]. The primary advantage of antenna arrays lies in their capability for dynamic beamforming. By carefully adjusting the amplitude and phase of signals feeding each radiating element, the combined field can be shaped and steered electronically, allowing the antenna to track moving users or targets without mechanical rotation [7]. For millimetre-wave 5G systems, antennas must be extremely compact while maintaining high gain and narrow half-power beam widths (HPBW). The array proposed in this work leverages phase-controlled multiport excitation to steer the beam and manage the radiation direction effectively, an essential requirement given the propagation challenges associated with high frequency signals [8-9].

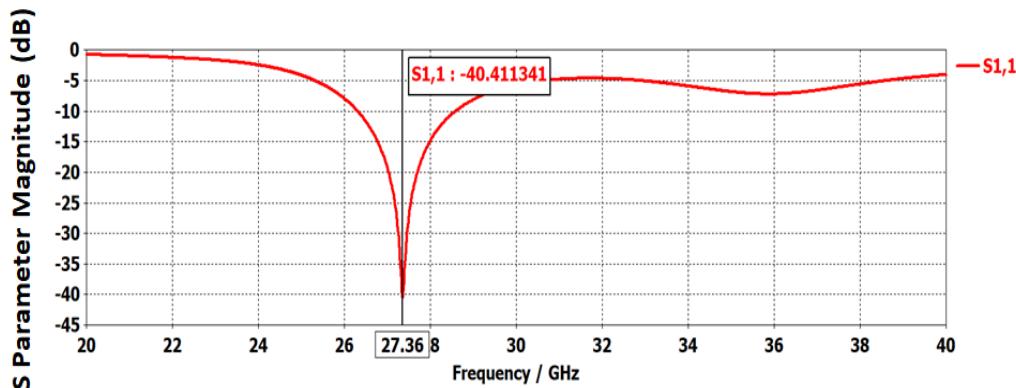
This research aims to design an efficient antenna operating in the 28 GHz band allocated for 5G GSM services by ITU standards. The proposed structure utilizes low-loss substrate materials and maintains a compact footprint while supporting multi-directional signal reception through a multiport configuration. Phase differences across the feeding ports are optimized to generate a narrow, highly directive beam with a gain exceeding 10 dB. The fundamental radiating element is a compact  $2.95 \times 2.95 \text{ mm}^2$  patch tailored for 28 GHz operation. Due to its small size, a transformer-coupled feed mechanism is adopted to achieve efficient excitation and support multi-beam radiation. The initial dual-port orthogonal transformer-fed design is expanded into two-element and four-element configurations, culminating in a five-element array with eight input ports. The central element is parasitically excited to enhance boresight directivity. Beam orientation is precisely controlled by adjusting input phase differences, enabling highly directive radiation at desired angles. Comprehensive optimization and experimental validation including comparison of measured and simulated return loss confirms the robustness and effectiveness of the proposed antenna system.

## 2. Single patch antenna design and analysis



**Figure 1.** Single patch antenna with matching transformer and feed line

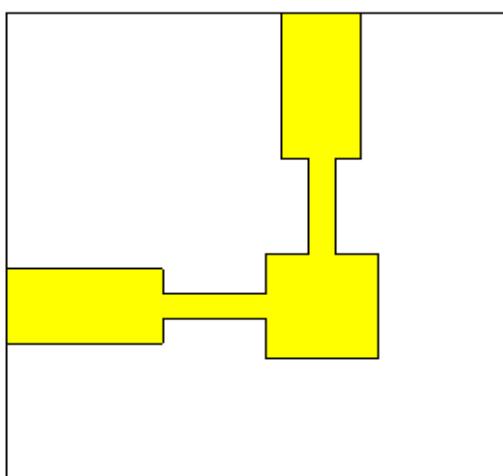
The design of the antenna array begins with the creation of a single radiating patch element, which subsequently serves as the foundational unit for constructing the complete array. This patch is excited through a microstrip feed line incorporating a quarter-wave impedance transformer to ensure proper matching. As shown in figure 1, the radiating element is a square patch measuring 2.95 mm  $\times$  2.95 mm. The antenna is fabricated on a Rogers RT5880LZ substrate, selected for its low permittivity value of 1.96 and thickness of 1.27 mm, and is optimized to operate at the target frequency of 28 GHz. The quarter-wave transformer is dimensioned at 0.72 mm  $\times$  2.7 mm, whereas the feed line length and width are 2.1 mm  $\times$  4.1 mm. Full-wave simulations of the single-port configuration were performed, and the resulting reflection coefficient ( $S_{11}$ ), shown in figure 2, indicates an initial resonance at 27.36 GHz with a minimum  $S_{11}$  of  $-40.41$  dB, demonstrating excellent impedance matching.



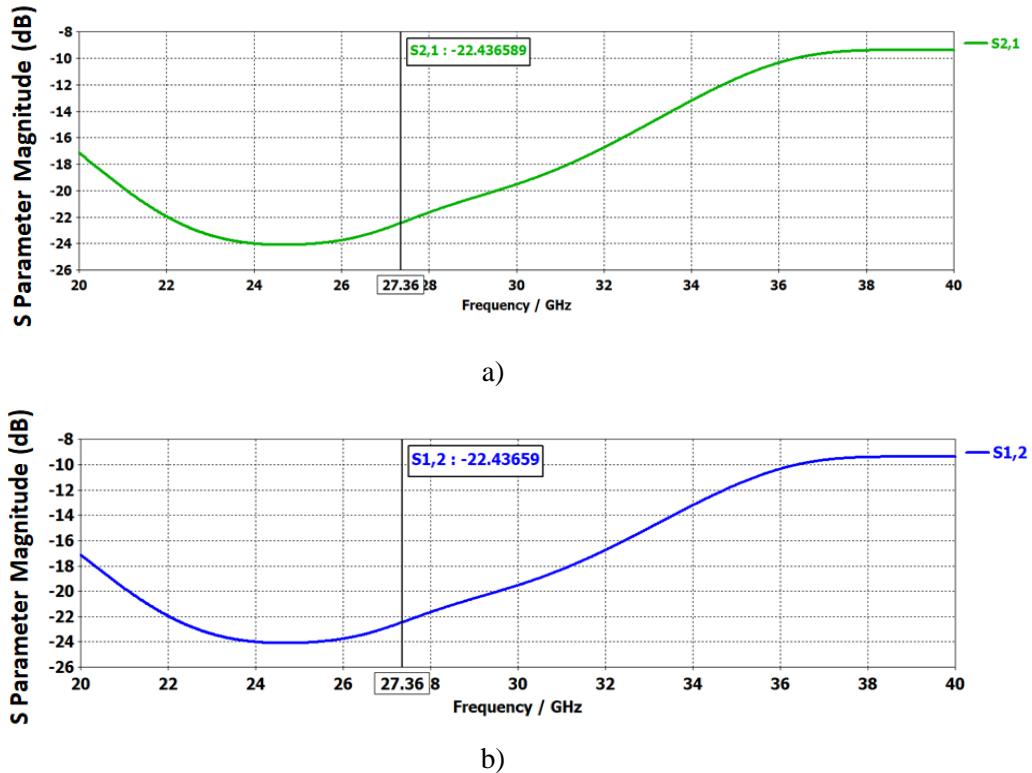
**Figure 2.**  $S_{11}$ (dB) vs frequency for single patch antenna

### 3. Multiple ports feeding

To enable the formation of multiple radiation beams, an optimized orthogonal feeding arrangement is employed. In this configuration, the patch is energized through two perpendicular feed ports, which allows the surface current distribution to be altered depending on the active port, as illustrated in figure 3. Ensuring low mutual coupling between these orthogonal ports is crucial for independent beam control. As shown in figure 4, the simulated  $S_{12}$  and  $S_{21}$  responses are identical, with both achieving  $-22.43$  dB at 27.36 GHz. This low coupling level results from the orthogonal placement of the feeds, confirming that the two ports function independently with minimal interference.



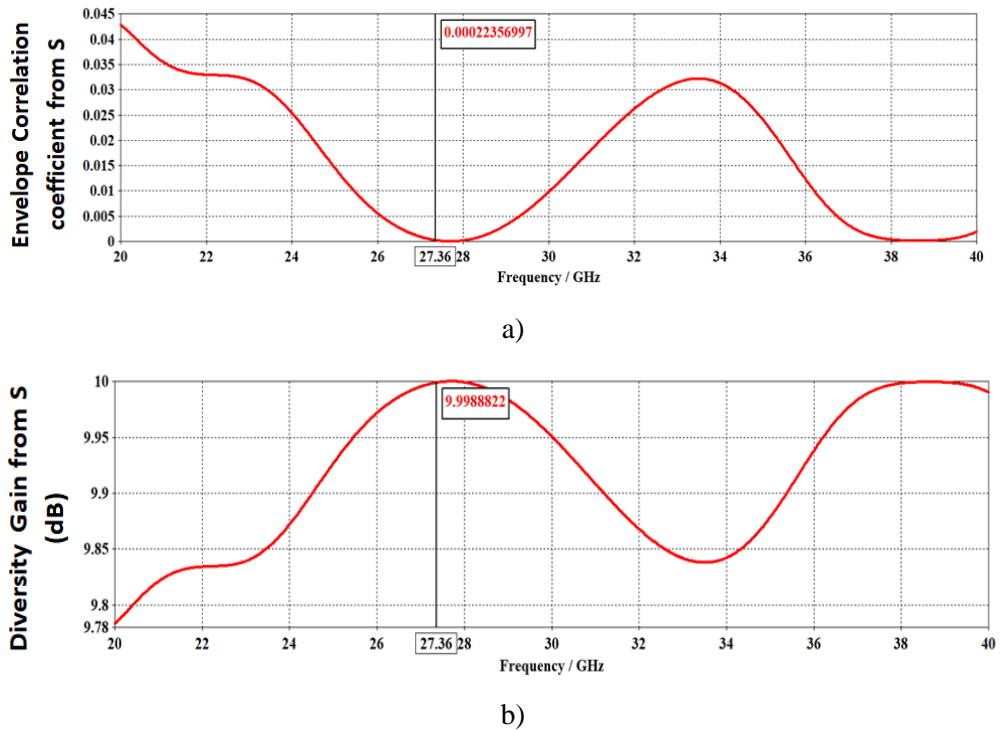
**Figure 3.** Multiple port feeding



**Figure 4.** Graphs for a)  $S_{21}$  b)  $S_{12}$  for multiple port antenna

$$\rho = \frac{(S_{11}^* S_{12} + S_{21}^* S_{22})^2}{(1 - |S_{11}|^2 - |S_{21}|^2)(1 - |S_{22}|^2 - |S_{12}|^2)} \quad (1)$$

$$DG(dB) = 10 * \sqrt{1 - (0.99 * \rho)^2} \quad (2)$$

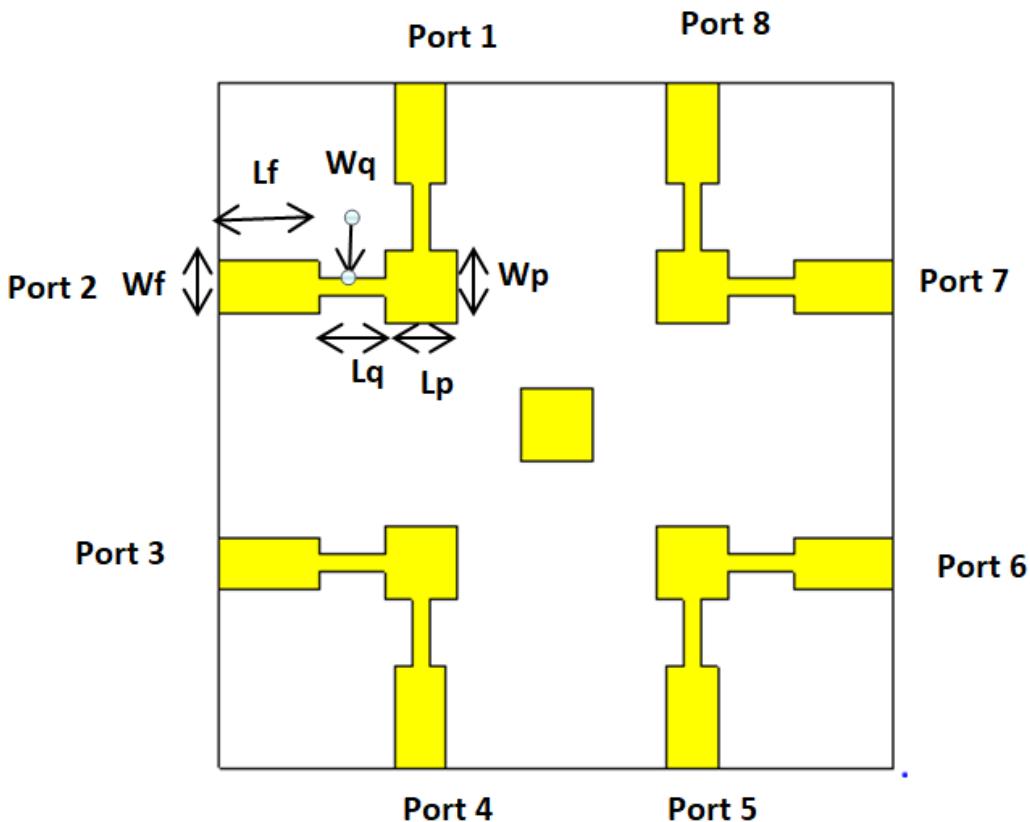


**Figure 5.** Graphs for a) Correlation Coefficient b) Diversity Gain with frequency for multiport antenna

The correlation coefficient ( $\rho$ ) and diversity gain between the two feeding ports are evaluated using the expressions provided in equations 1 and 2. Using these formulations, the calculated correlation coefficient is 0, while the diversity gain is 9.99 dB. The simulated variation of these parameters across the operating band is presented in figure 5. At the resonance frequency of 27.36 GHz, the correlation coefficient remains extremely low, and the diversity gain approaches the ideal value of 10 dB. These results confirm that the two ports exhibit minimal correlation and operate effectively as independent channels.

#### 4. Final antenna array with parasitic patch at the centre

The finalized configuration of the five-element antenna array, incorporating a centrally positioned patch, is presented in figure 6. In this arrangement, the central element is not directly fed but instead energized through parasitic coupling from the surrounding four patches, enhancing the strength and focus of the main radiation beam. The overall antenna structure maintains a compact footprint of  $27.7 \times 27.7 \times 1.27$  mm<sup>3</sup>, making it significantly smaller than many previously reported designs. All key geometric and structural parameters of the array are summarized in table 1.

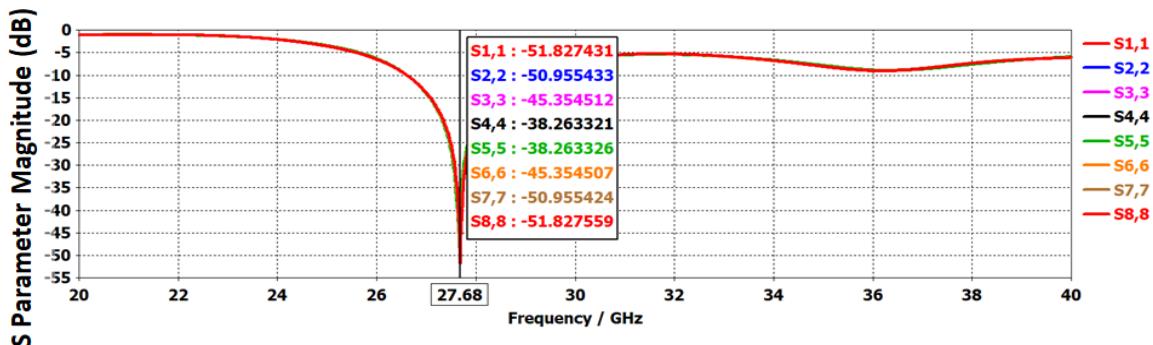


**Figure 6.** Five element antenna arrays

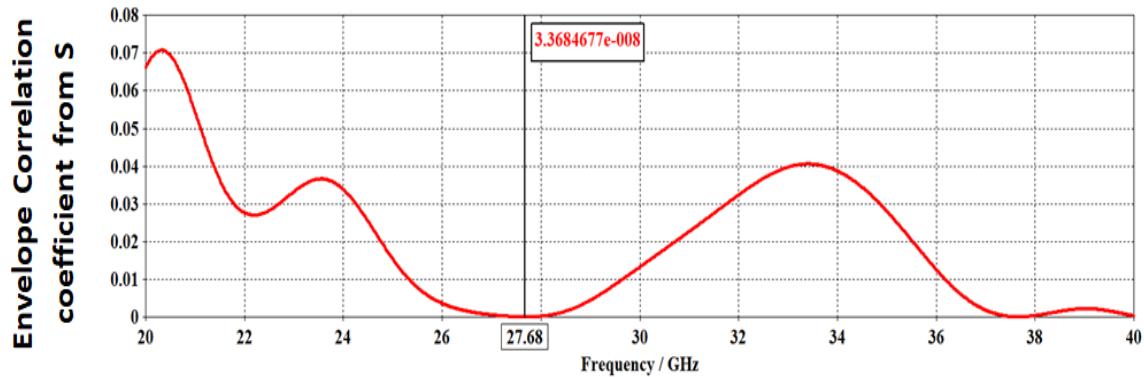
All dimensions in mm	
Wf	2.1
Lf	4.1
Wq	0.72
Lq	2.7
Wp	2.95
Lp	2.95

**Table 1.** Dimensions of microstrip antenna array

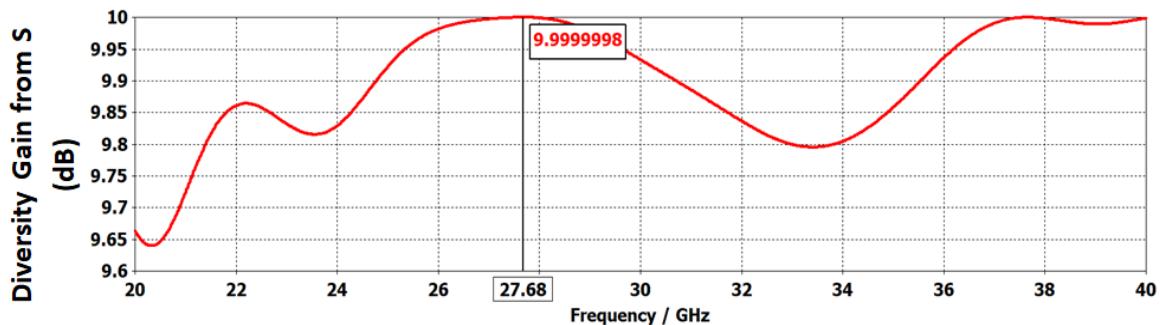
The simulated performance of the proposed antenna array in terms of return loss and correlation coefficient is presented in figure 7 and figure 8, respectively. At the resonant frequency of 27.68 GHz, the array achieves an excellent return loss of  $-51.8$  dB, while the correlation coefficient between the feeding ports is evaluated as zero, indicating complete independence between the ports. The corresponding diversity gain at this frequency is 9.99 dB, as depicted in figure 9. The surface current distribution illustrated in figure 10 shows that when all ports are excited with an identical phase ( $0^\circ$  shift), the currents on the radiating patches are directed toward the central element, demonstrating effective multi-port excitation and confirming the intended array behavior.



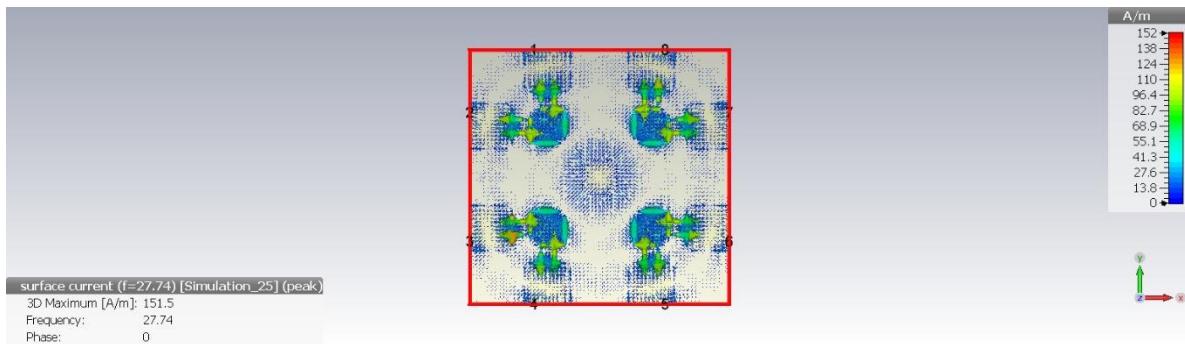
**Figure 7.** S parameter (dB) vs Frequency graph for the antenna array



**Figure 8.** Correlation Coefficient variation with frequency for the array



**Figure 9.** Diversity Gain variation with frequency for the array

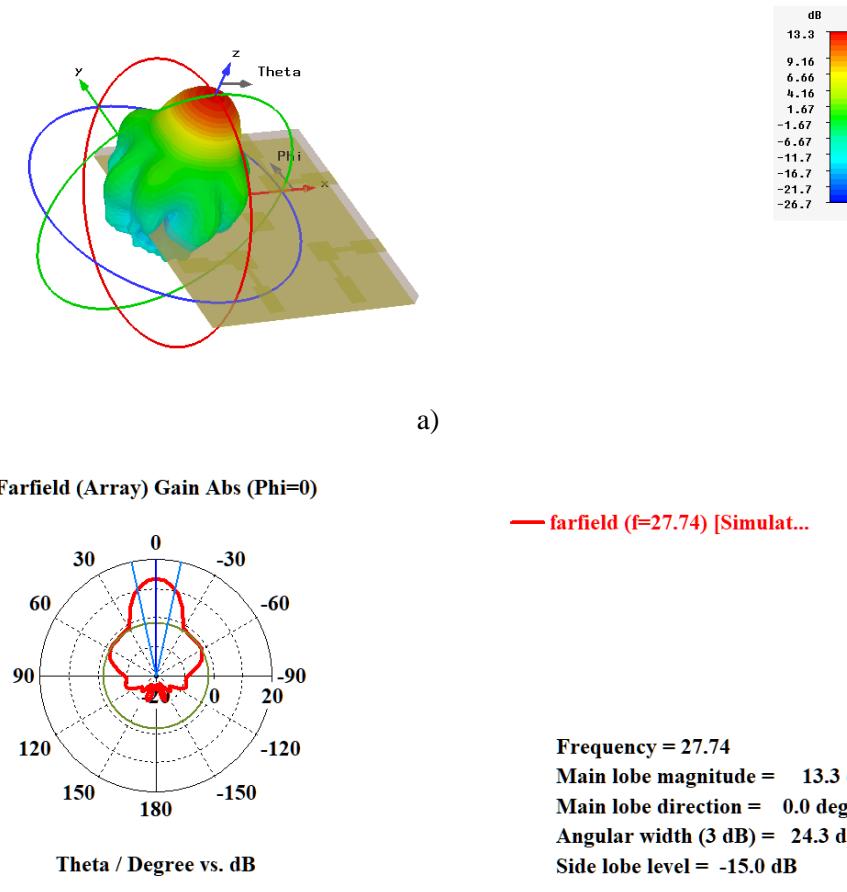


**Figure 10.** Current distribution on five-element array

## 5. Multiple beam pattern

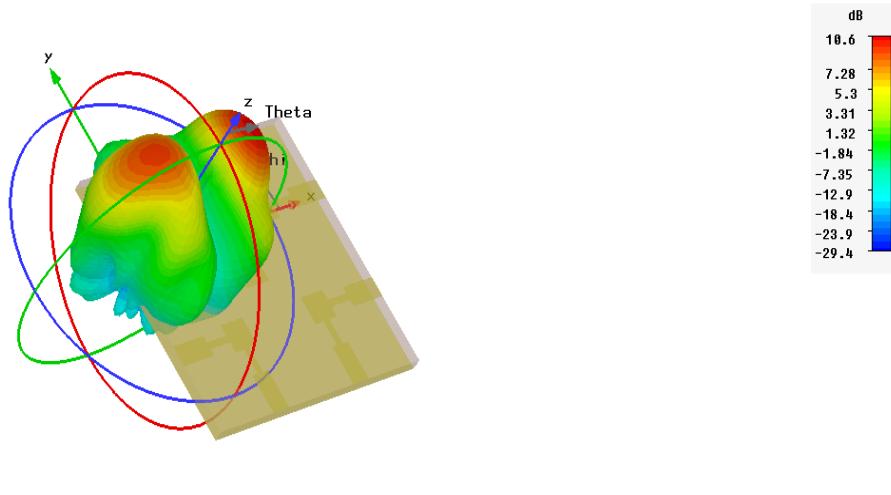
The optimized antenna array features five radiating patches powered through eight feeding ports, allowing flexible beam formation through controlled phase variation among the input signals. By adjusting the phase differences applied to these ports, the array can generate multiple radiation beams oriented in different directions. Depending on the selected phase configuration, the primary radiation can split into two, three, four, or even five distinct beams. The optimized outcomes for each case are illustrated using 3D radiation patterns and corresponding polar plots.

**Case 1:** In the first scenario, where a phase increment of  $45^\circ$  is applied across the feeding ports, the array produces a single, well-defined radiation beam, as depicted in figure 11.



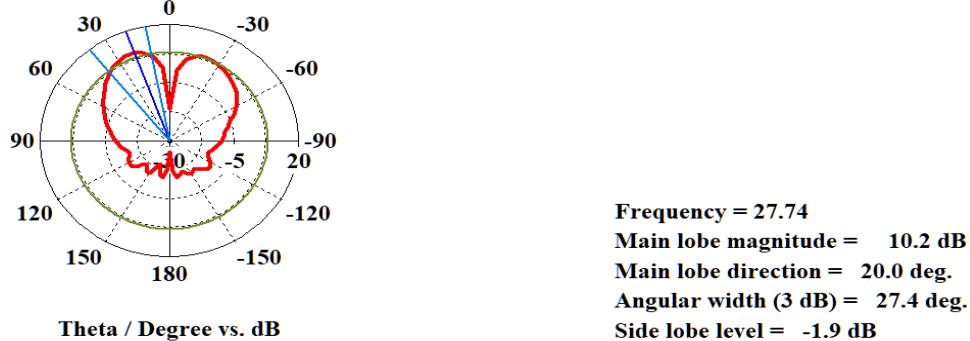
**Figure 11.** Radiation pattern for single beam (Case 1) a) 3D plot b) Polar plot

**Case 2:** When a phase shift of  $110^\circ$  is applied sequentially across the feeding ports, the antenna array generates two distinct radiation beams. This phase configuration effectively splits the main lobe into a dual-beam pattern, as illustrated in figure 12, demonstrating the array's ability to support multi-directional beam formation.



a)

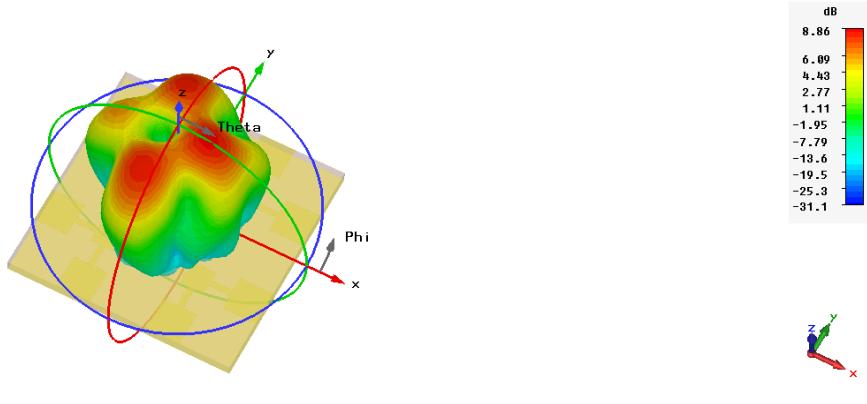
Farfield (Array) Gain Abs (Phi=0)  
— farfield (f=27.74) [Simulat...



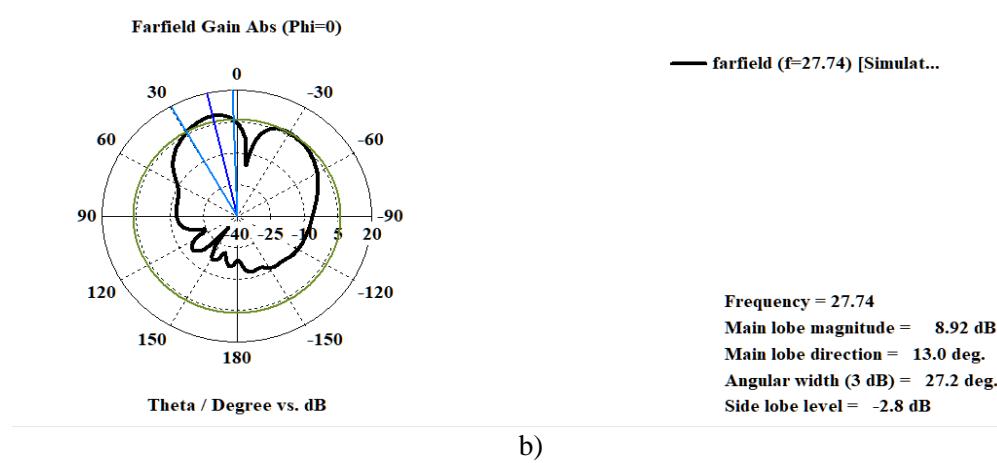
b)

Figure 12. Radiation pattern for two beams (Case 2) a) 3D plot b) Polar plot

**Case 3:** In this configuration, a  $180^\circ$  phase shift is introduced between the two ports of each individual patch, while maintaining an out-of-phase condition of  $45^\circ$  between adjacent patches. Under this combined phase arrangement, the antenna array produces three distinct radiation beams, as illustrated in figure 13. This result confirms the array's capability to achieve controlled multi-beam operation through tailored intra-element and inter-element phase variations.

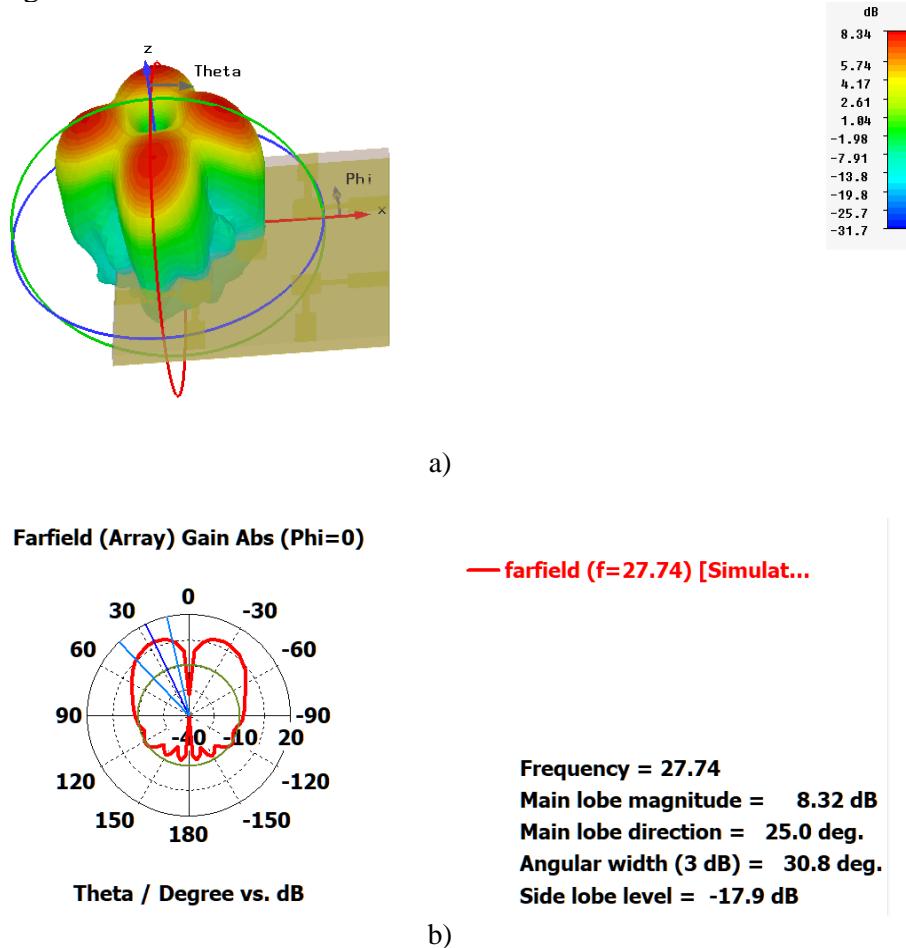


a)



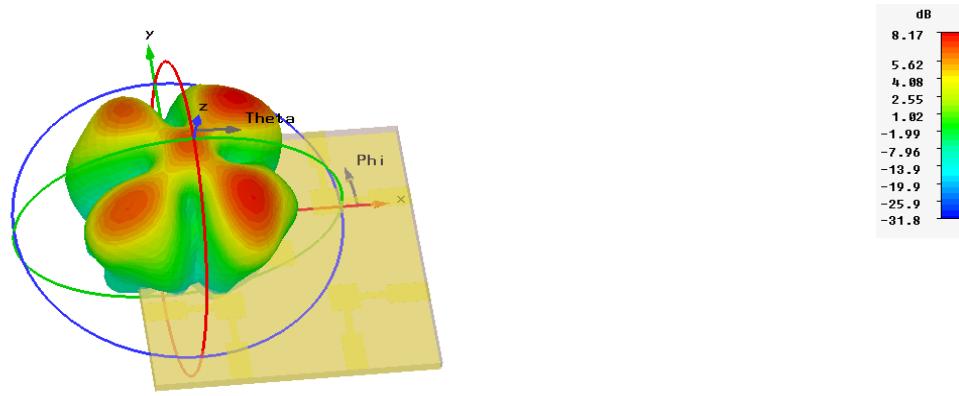
**Figure 13.** Radiation pattern for three beams (Case 3) a) 3D plot b) Polar plot

**Case 4:** When a uniform phase shift of  $90^\circ$  is applied across all feeding ports, the antenna array produces four distinct radiation beams. This phase configuration enables the main lobe to split into four symmetrical directions, forming a well-defined quad-beam pattern, as illustrated in figure 14.



**Figure 14.** Radiation pattern for four beams (Case 4) a) 3D plot b) Polar plot

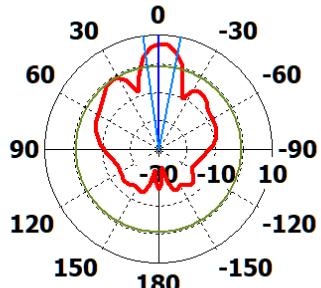
**Case 5:** By applying a phase shift of  $140^\circ$  sequentially across the feeding ports, the antenna array produces five distinct radiation beams. This configuration results in a well-defined five-beam pattern, as depicted in figure 15, demonstrating the array's capability to support wide-angle multi-beam radiation through appropriate phase control.



a)

**Farfield (Array) Gain Abs (Phi=0)**

— farfield (f=27.74) [Simulat...



Theta / Degree vs. dB

**Frequency = 27.74**  
**Main lobe magnitude = 6.79 dB**  
**Main lobe direction = 0.0 deg.**  
**Angular width (3 dB) = 19.0 deg.**  
**Side lobe level = -7.4 dB**

b)

**Figure 15.** Radiation pattern for five beams (Case 5) a) 3D plot b) Polar plot

## 6. Fabricated design and results

The finalized five-element patch array was fabricated on a Rogers RT5880LZ substrate ( $\epsilon_r = 1.96$ ) and experimentally characterized using a Rohde & Schwarz ZVA 40 Vector Network Analyzer, as shown in figure 16. The measured  $S_{11}$  response, presented in figure 17, exhibits a minimum at 27.8 GHz. A comparison of simulated and measured return loss results is illustrated in figure 18. The simulated design achieves an optimal resonance at 27.6 GHz, while the fabricated prototype resonates closely at 27.8 GHz. This close agreement between both results confirms the accuracy of the design approach and validates the overall performance of the proposed antenna array.



a)

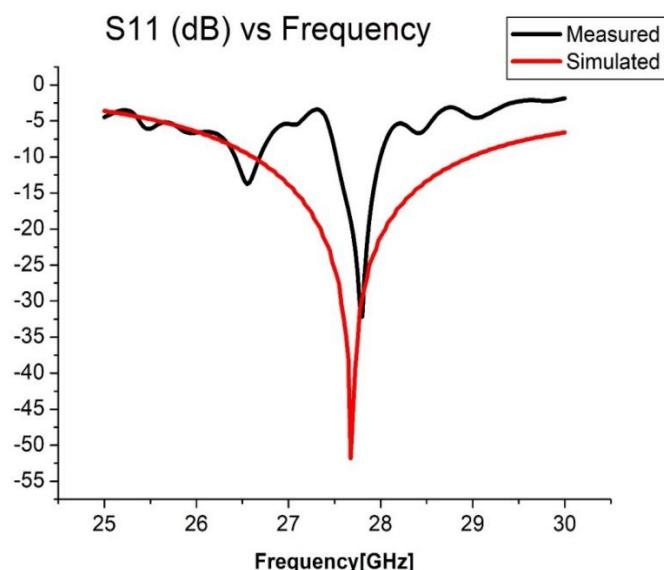


b)

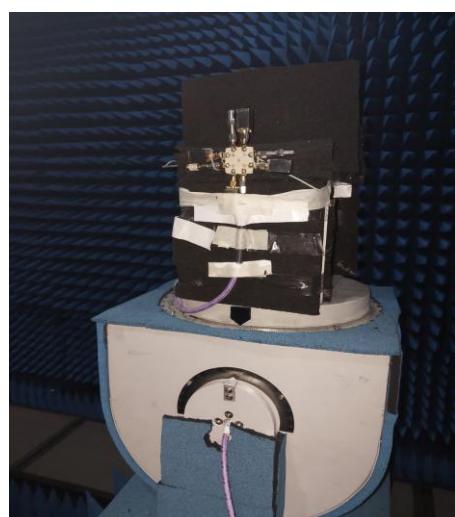
**Figure 16.** a) Fabricated 5 patch antenna array and b) measurement on VNA



**Figure 17.** Screenshot of VNA for measurement of fabricated design



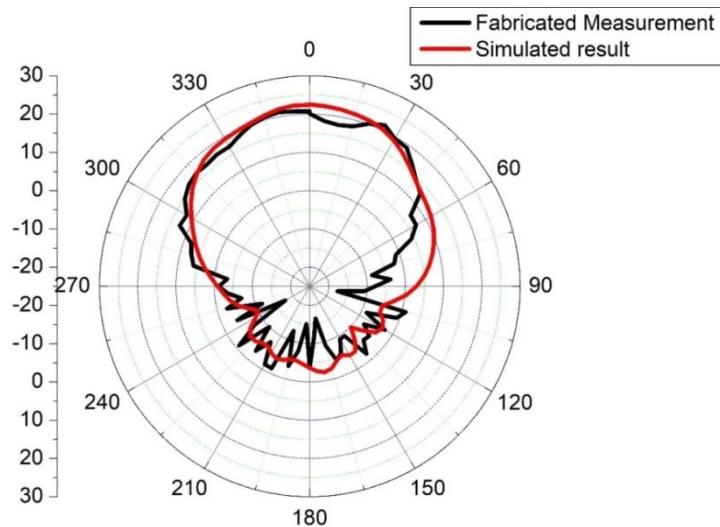
**Figure 18.** Comparison of simulated and measured results



**Figure 19.** Radiation pattern measurement of antenna array in anechoic chamber

The finalized eight-port configuration shown in figure 6 was fabricated as illustrated in figure 16. For measurement, only a single port (Port 1) was excited while all remaining ports were terminated with matched loads. The radiation pattern of this setup was evaluated inside an anechoic chamber capable of supporting measurements up to 70 GHz, as depicted in figure 19.

Figure 20 presents the comparison between the simulated and measured radiation patterns. The measurements were recorded in 5 degree increments over the full 0–360° range at the operating frequency of 27.7 GHz. The resulting curves show a close agreement between the measured data and the simulated predictions.



**Figure 20.** Comparison of Radiation patterns of fabricated design with simulated result

## 7. Conclusion

This work has demonstrated the complete development and verification of a compact multi-beam microstrip patch array featuring five radiating elements and eight excitation ports. Through the application of tailored phase configurations at the inputs, the array successfully produces one to five independent beams, enabling flexible coverage and the capability to receive signals from diverse propagation paths which is an essential feature for modern 5G systems. While the formation of multiple beams leads to a modest reduction in gain due to power being shared among various directions, the antenna's overall performance remains robust and well suited for realistic multi-path wireless scenarios.

### Authors' Declaration

The authors declare that there is no conflict of interest regarding the publication of this paper.

### Authors' Contribution Statement

Conceptualization, methodology, investigation, writing (original draft), H.N.; writing (review and editing of the manuscript), H.N, MM; visualization, supervision, M.M

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